

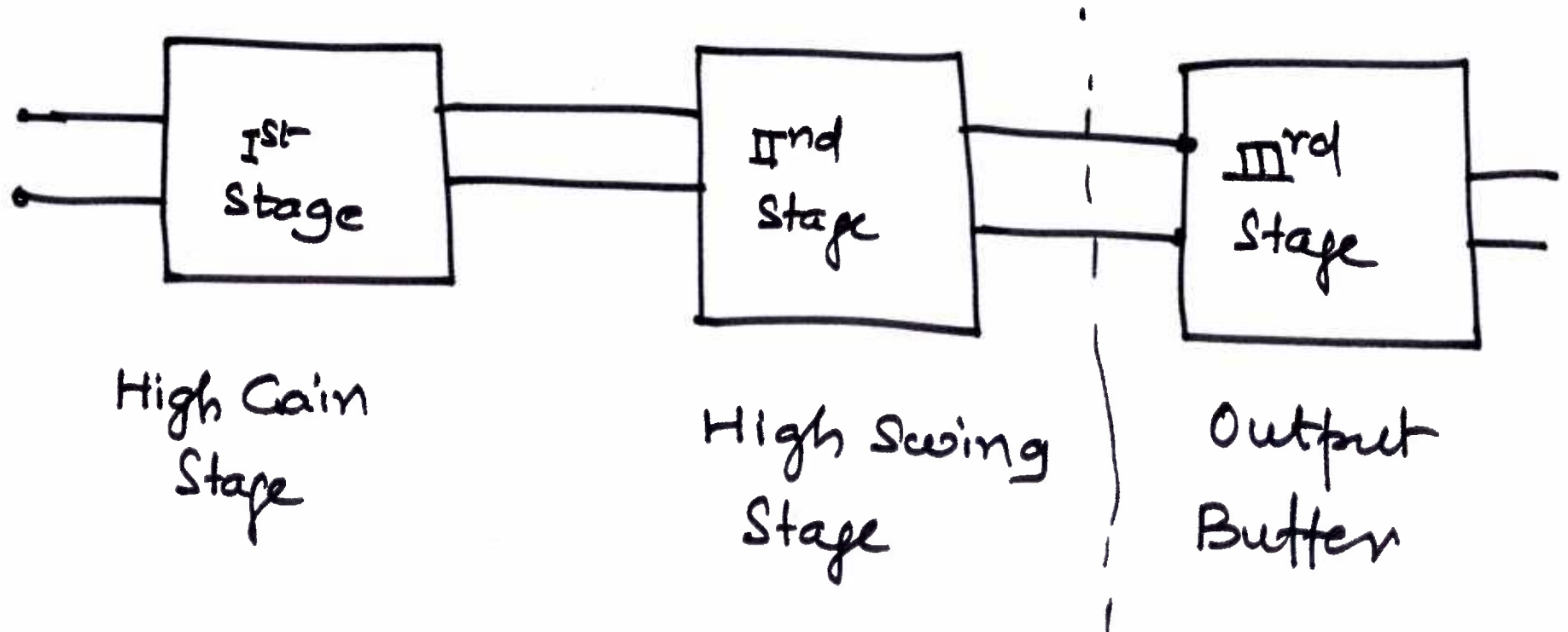
CMOS OPAMP : Two Stage OPAMP

Typically Double-ended CMOS uses Current Source biasing, while Single-ended OPAMP uses Current Mirror Biasing



CDEEP
IIT Bombay

EE 618 L ___ / Slide ___

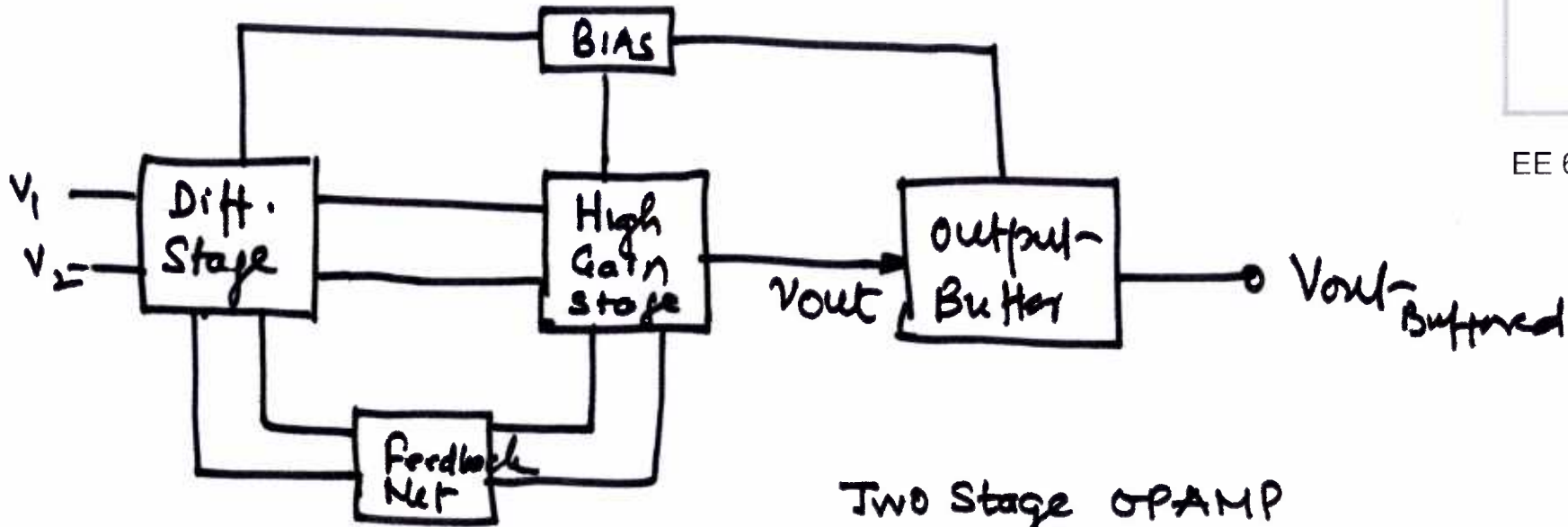


Block Diagram of OPAMP

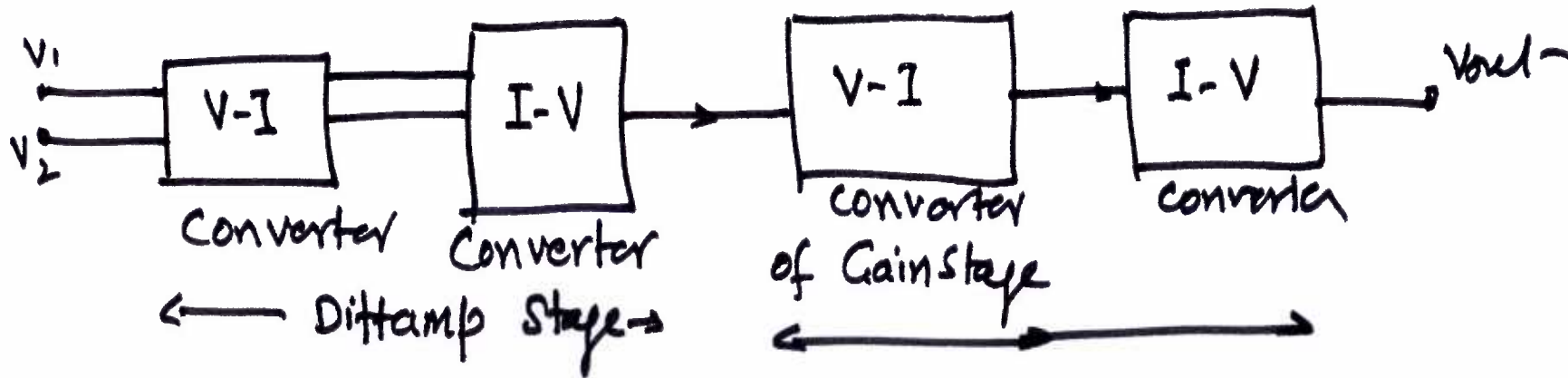


CDEEP
IIT Bombay

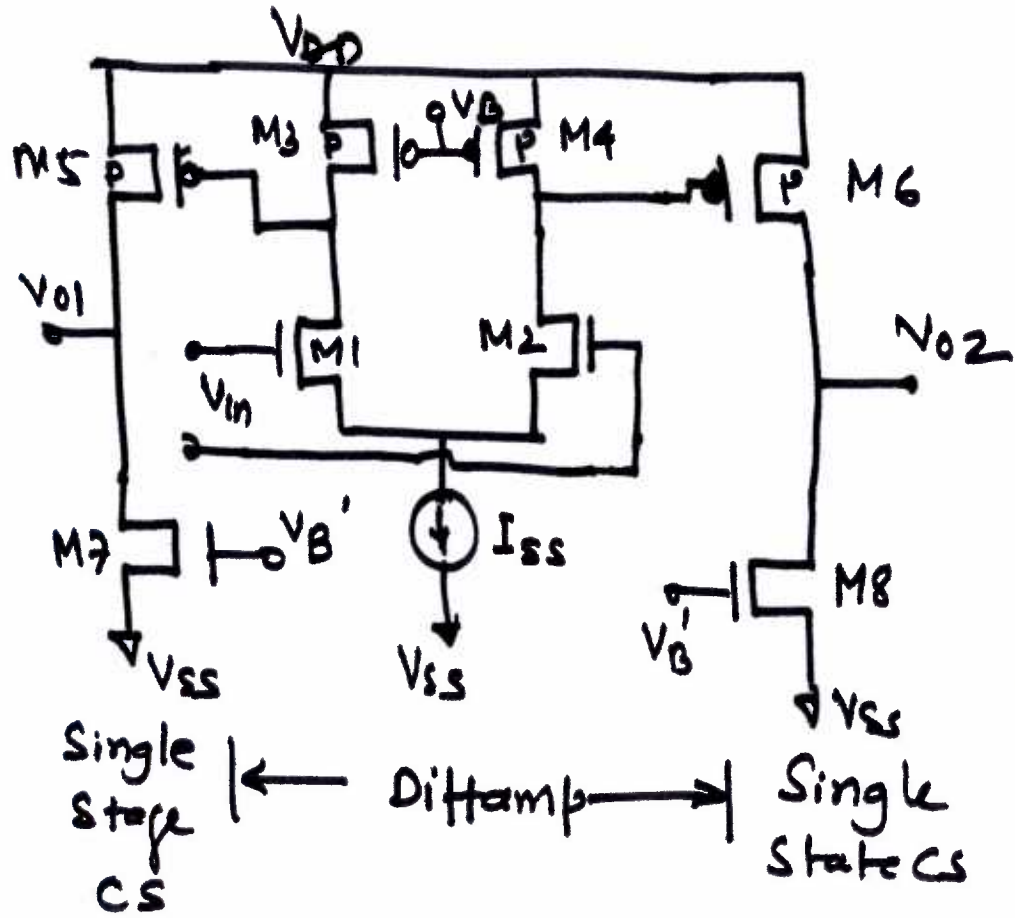
EE 618 L ___ / Slide ___



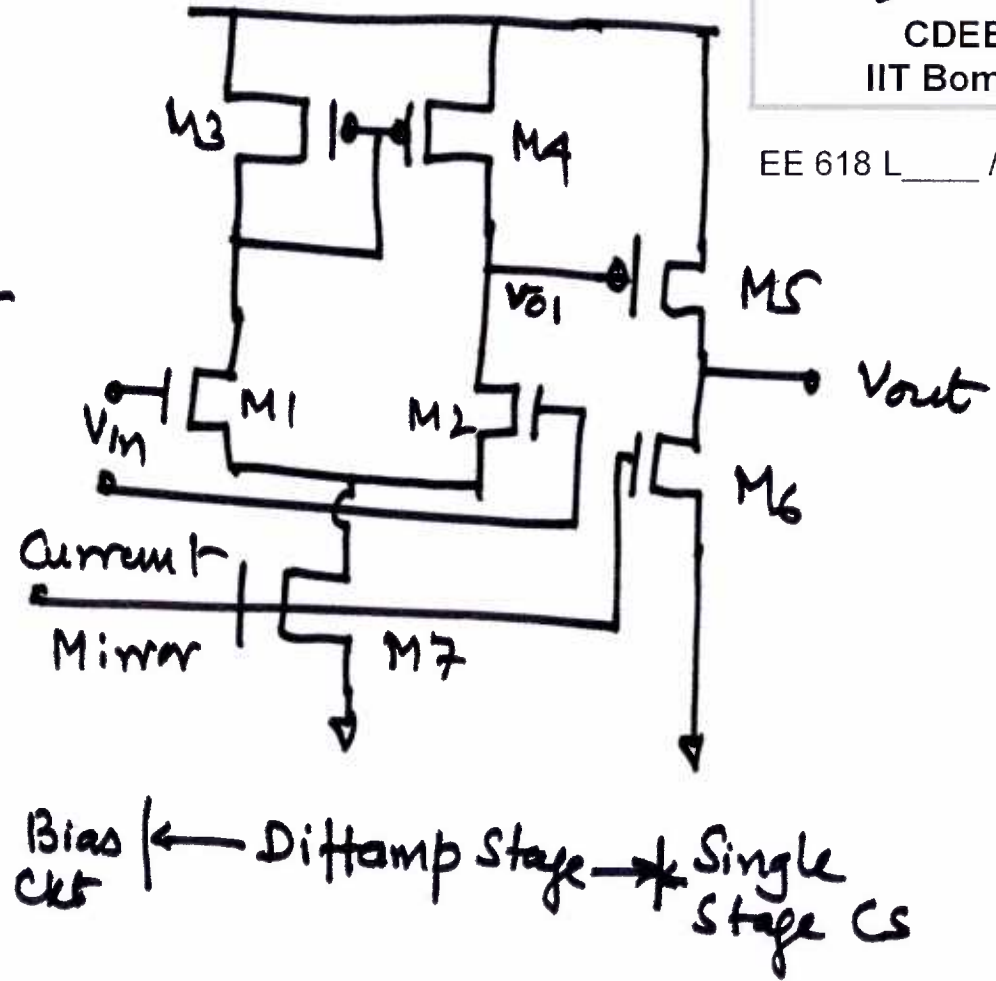
Two Stage OPAMP



Two Stage Double-ended OPAMP



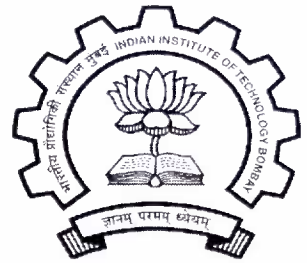
Two Stage Single-ended OPAMP



CDEEP
IIT Bombay

EE 618 L / Slide

For double ended cas



CDEEP
IIT Bombay

EE 618 L / Slide

$$V_{o1} = -g_{m1} (r_{o1} || r_{o3}) V_{in1}$$

$$V_{o2} = -g_{m2} (r_{o2} || r_{o4}) (-V_{o1})$$

$$V_{o2} = +g_1 g_{m5} (r_{o1} || r_{o3}) (r_{o2} || r_{o4}) V_{in}$$

$$\begin{aligned} \therefore A_{V1} = A_{V2} &= \frac{V_{o1}}{V_{in}} = \frac{V_{o2}}{V_{in}} = g_{m1} g_{m5} (r_{o1} || r_{o3}) (r_{o2} || r_{o4}) \\ &= g_{m2} g_{m6} (r_{o6} || r_{o8}) (r_{o2} || r_{o4}) \end{aligned}$$

$$r_{o1} = \frac{1}{\lambda_1 I_{SS}/2} \quad \text{and} \quad r_{o3} = \frac{1}{\lambda_3 I_{SS}/2}, \quad r_{o5} = \frac{1}{\lambda_5 I_{SS}/2}$$
$$r_{o2} = \frac{1}{\lambda_2 I_{SS}/2}, \quad r_{o4} = \frac{1}{\lambda_4 I_{SS}/2}; \quad r_{o6} = \frac{1}{\lambda_6 I_{SS}}$$

Here for single ended case

$$A_V = A_{V1} \cdot A_{V2}$$

$$A_{V1} = \frac{V_{o1}}{V_{in}} = -g_{m1} \cdot (r_{o2} \parallel r_{o4})$$

However $V_{o1} = V_{in2}$ for CS Amplifier

$$\therefore A_{V2} = \frac{V_{out}}{V_{o1}} = -g_{m5} \cdot (r_{o5} \parallel r_{o6})$$

$$\therefore A_V = +g_{m1} g_{m5} (r_{o2} \parallel r_{o4}) (r_{o5} \parallel r_{o6})$$

$$g_{m1} = \sqrt{2\beta_1 I_{DS1}} = \sqrt{2\beta_2 I_{DS2}} = g_{m2}$$

$$I_{DS1} = I_{DS2} = \frac{I_{SS}}{2} \quad \therefore g_{m1} = \sqrt{\beta_1 I_{SS}} = g_{m2}$$



CDEEP
IIT Bombay

EE 618 L ____ / Slide ____

If Bias current is chosen as $20 \mu A$

$$\therefore I_{SS} = 20 \mu A, \quad \therefore \frac{I_{SS}}{2} = 10 \mu A.$$

$$\therefore I_{DS1} = I_{DS3} = I_{DS2} = I_{DS4} = 10 \mu A \quad (\text{at } V_{id} = 0)$$

$$\therefore g_{m1} = \sqrt{2\beta_1 \frac{I_{SS}}{2}}$$

$$r_{o1} = r_{o2} = \frac{1}{\lambda_2 I_{SS}/2}$$

$$r_{o4} = \frac{1}{\lambda_4 I_{SS}/2}$$

$$\frac{I_{SS}}{2} = \frac{\beta_1}{2} \cdot (V_{GS1} - V_{TN})^2$$

We choose Technology of 5μ .

$$\beta_n' = 50 \mu A/V^2$$

$$\beta_p' = 16 \mu A/V^2$$

$$\lambda_1 = \lambda_2 = 0.06$$

$$\lambda_3 = \lambda_4 = 0.05/V$$

$$= \frac{1}{\lambda_2 I_{SS}/2}$$

$$\therefore r_{o2} || r_{o4} = \frac{1}{(\lambda_2 + \lambda_4) I_{SS}/2}$$

$$\text{or } \sqrt{I_{SS}} = \sqrt{\frac{\beta_1}{2}} (V_{GS1} - V_{TN})$$



CDEEP
IIT Bombay

EE 618 L ____ / Slide ____



CDEEP
IIT Bombay

EE 618 L ____ / Slide ____

Using Data from Boyce, Baker, Li's book,

For $V_{OV} = 0.37 \text{ V}$, $V_{Tn} = 0.83 \text{ V}$, $V_{Tp} = -0.9 \text{ V}$

$$V_{GS1} = V_{GS2} = 0.83 + 0.37 \text{ V} = 1.2 \text{ V}$$

For $I_{SS} = 20 \mu\text{A}$, or $I_{DS1} = I_{DS2} = +I_{SS}/2$

$$\text{We get } \left(\frac{W}{L}\right)_n = \frac{15}{5} \text{ and } \left(\frac{W}{L}\right)_p = \frac{70}{5}$$

Then Open Loop Gain of Two Stage OPAMP can be

found = $|A_{OL}|$

$$= \sqrt{2 \times 50 \times 10^{-6} \left(\frac{15}{5}\right) 10 \times 10^{-6}} \times \sqrt{2 \times 16 \times 10^{-6} \left(\frac{70}{5}\right) 10 \times 10^{-6}}$$

$\leftarrow g_{m1} \quad \rightarrow \quad \leftarrow g_{m5} \quad \rightarrow$

$$* \left[\frac{1}{(0.06 + 0.06) \times 10 \times 10^{-6}} \right]^2$$

$(r_{O2} || r_{O4}) (r_{O5} || r_{O6})$

Then $|AOL| \approx 2500 \text{ V/V}$

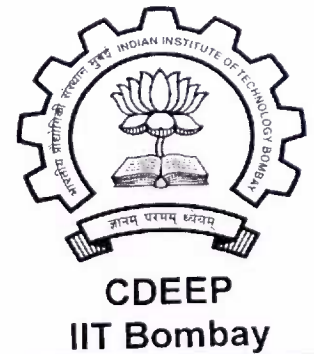
It is obvious from AOL expression that

$$AOL \propto \frac{1}{I_{SS}}$$

Hence increase of I_{SS} may improve Bandwidth but decrease AOL, or Vice-a-Versa.

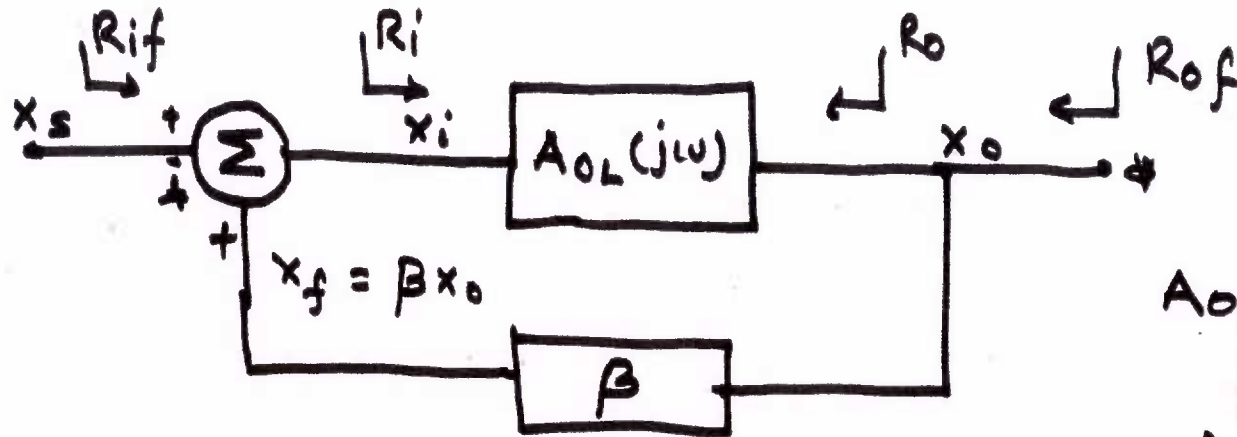
In real life OPAMP, one wishes to improve the stability of Gain against any variations,

So we wish to visit 'Feedback & Stability' issue in Brief, so that we can achieve desired Gain, Bandwidth and other specs, and also System is Stable.



Feedback

CDEEP
IIT Bombay



$A_{OL}(j\omega)$ = Open Loop Gain

β = Feedback Factor

$A_{CL}(j\omega)$ = Closed Loop Gain

General Feedback System

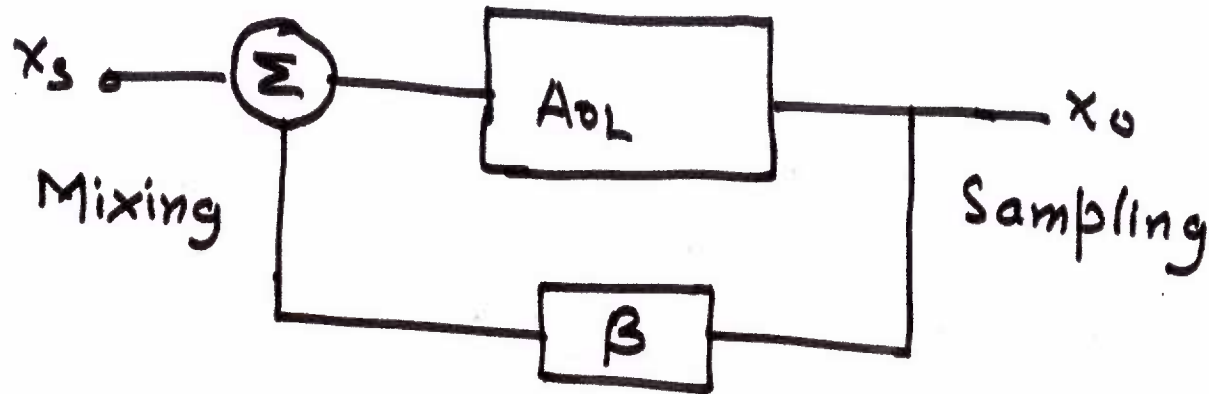
$$x_o = A_{OL}(j\omega) \cdot x_i = A_{OL}(j\omega) [x_s - x_f]$$

Here

$$x_i = x_s - x_f \quad \text{And} \quad x_f = \beta x_o$$

$$\therefore x_o = A_{OL} x_s - A_{OL} \beta x_o \quad \therefore A_{CL}(j\omega) = \frac{x_o}{x_s} = \frac{A_{OL}}{1 + A_{OL} \beta}$$

Definitions



$A_{OL} = A_0 = \text{Gain without feedback}$

$A_{CL} = \text{Closed Loop Gain}$

$A_{OL}\beta = \text{Loop Gain}$

$1 + A_{OL}\beta = \text{Amount of Feedback}$

$A_{OL}\beta = \text{Return Ratio}$



CDEEP
IIT Bombay